

# EEE4001F: Digital Signal Processing

Class Test 1

11 March 2016

## SOLUTIONS

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Name:

Student number:

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### Information

- The test is closed-book.
  - This test has *four* questions, totalling 20 marks.
  - Answer *all* the questions.
  - You have 45 minutes.
  - An information sheet is attached.
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1. (5 marks) A discrete-time system is governed by the following relation:

$$y[n] = \sum_{k=0}^2 x[n-k] + x[0].$$

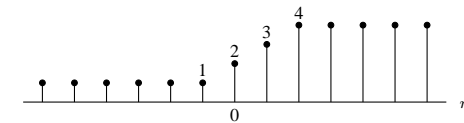
- Find the output when the input is  $x_1[n] = u[n]$ .
- Find the output when the input is  $x_2[n] = u[n-1]$ .
- Is the system time invariant?

The input-output relationship can be written as

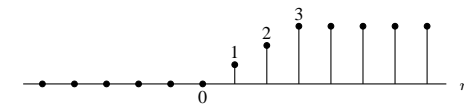
$$y[n] = x[n] + x[n-1] + x[n-2] + x[0],$$

and for each  $n$  the required values can be found by direct substitution.

(a) Response for  $x_1[n] = u[n]$  is  $y_1[n]$  below:



(b) Response for  $x_2[n] = u[n-1]$  is  $y_2[n]$  below:



(c) Since  $x_2[n] = x_1[n]$  but  $y_2[n] \neq y_1[n]$  the system is not time invariant.

2. (5 marks) Determine the DTFT of the sequence  $x[n] = \alpha^n u[-n - 1]$  for  $|\alpha| > 1$ .

Can approach the problem from first principles:

$$\begin{aligned} X(e^{j\omega}) &= \sum_{n=-\infty}^{\infty} x[n]e^{-j\omega n} = \sum_{n=-\infty}^{\infty} \alpha^n u[-n - 1]e^{-j\omega n} = \sum_{n=-\infty}^{-1} \alpha^n e^{-j\omega n} \\ &= \sum_{n=1}^{\infty} \alpha^{-n} e^{j\omega n} = -1 + \sum_{n=0}^{\infty} (\alpha^{-1} e^{j\omega})^n. \end{aligned}$$

Since we know that  $|\alpha| > 1$  we have  $|\alpha^{-1} e^{j\omega}| < 1$ , so this infinite series converges to

$$X(e^{j\omega}) = -1 + \frac{1}{1 - \alpha^{-1} e^{j\omega}} = \frac{\alpha^{-1} e^{j\omega}}{1 - \alpha^{-1} e^{j\omega}}$$

Alternatively one could use the given z-transform pair

$$-\alpha^n u[-n - 1] \xleftrightarrow{\mathcal{Z}} \frac{1}{1 - \alpha z^{-1}} \quad |z| < |\alpha|$$

to obtain

$$\alpha^n u[-n - 1] \xleftrightarrow{\mathcal{Z}} -\frac{1}{1 - \alpha z^{-1}} \quad |z| < |\alpha|.$$

Thus

$$H(z) = \frac{-1}{1 - \alpha z^{-1}} = \frac{z}{\alpha - z} = \frac{\alpha^{-1} z}{1 - \alpha^{-1} z} \quad \text{for } |z| < |\alpha|.$$

Now since  $|\alpha| > 1$  the ROC includes the unit circle and we can write

$$H(e^{j\omega}) = H(z)|_{z=e^{j\omega}} = \frac{\alpha^{-1} e^{j\omega}}{1 - \alpha^{-1} e^{j\omega}},$$

as before.

3. (5 marks) Suppose a sequence  $x[n]$  has DTFT  $X(e^{j\omega})$ . Find the time-domain inverses of each of the following:

(a)  $Y_1(e^{j\omega}) = 2X(e^{-j(\omega-\omega_0)})$ , and

(b)  $Y_2(e^{j\omega}) = 3e^{j4\omega} X(e^{j(\omega-\omega_0)})$ .

Express your answers in terms of  $x[n]$ .

(a) Applying the time reversal property to the given pair yields

$$x[-n] \xleftrightarrow{\mathcal{F}} X(e^{-j\omega}).$$

Applying frequency shifting to this gives the pair

$$e^{j\omega_0 n} x[-n] \xleftrightarrow{\mathcal{F}} X(e^{-j(\omega-\omega_0)}).$$

Finally from linearity

$$2e^{j\omega_0 n} x[-n] \xleftrightarrow{\mathcal{F}} 2X(e^{-j(\omega-\omega_0)}).$$

Thus  $y_1[n] = 2e^{j\omega_0 n} x[-n]$ .

(b) Using frequency shifting on the given pair yields

$$e^{j\omega_0 n} x[n] \xleftrightarrow{\mathcal{F}} X(e^{j(\omega-\omega_0)}).$$

Now time shift with  $n_d = -4$  on this pair gives

$$e^{j\omega_0(n+4)} x[n+4] \xleftrightarrow{\mathcal{F}} e^{j4\omega} X(e^{j(\omega-\omega_0)})$$

Using linearity provides the required result as  $y_2[n] = 3e^{j\omega_0(n+4)} x[n+4]$ .

4. (5 marks) Consider two discrete-time systems with the following impulse responses:

$$h_1[n] = \delta[n] - \delta[n - 1] \quad \text{and} \quad h_2[n] = u[n].$$

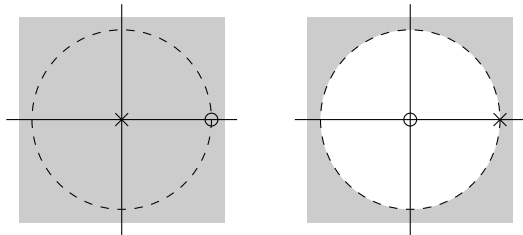
- Are the systems causal? Why?
- Using time-domain reasoning show that the systems are inverses of one another.
- Draw pole-zero plots of the system functions in each case.

- In both cases the impulse response satisfies  $h[n] = 0$  for  $n < 0$  so the systems are causal.
- The combined impulse response will be

$$\begin{aligned} h[n] &= h_1[n] * h_2[n] = (\delta[n] - \delta[n - 1]) * u[n] = \delta[n] * u[n] - \delta[n - 1] * u[n] \\ &= u[n] - u[n - 1] = \delta[n], \end{aligned}$$

so the output of the combined system will be identical to the input. Thus the systems are inverses of one another.

- The z-transform of  $h_1[n]$  is  $H_1(z) = 1 - z^{-1}$  (ROC all  $z$ ) and the transform of  $h_2[n]$  is  $H_2(z) = 1/(1 - z^{-1})$  (ROC  $|z| > 1$ ). The pole-zero plots are therefore as follows (left  $H_1(z)$  and right  $H_2(z)$ ):



## Fourier transform properties

Sequences $x[n], y[n]$	Transforms $X(e^{j\omega}), Y(e^{j\omega})$	Property
$ax[n] + by[n]$	$aX(e^{j\omega}) + bY(e^{j\omega})$	Linearity
$x[n - n_d]$	$e^{-j\omega n_d} X(e^{j\omega})$	Time shift
$e^{j\omega_0 n} x[n]$	$X(e^{j(\omega - \omega_0)})$	Frequency shift
$x[-n]$	$X(e^{-j\omega})$	Time reversal
$nx[n]$	$j \frac{dX(e^{j\omega})}{d\omega}$	Frequency diff.
$x[n] * y[n]$	$X(e^{-j\omega})Y(e^{-j\omega})$	Convolution
$x[n]y[n]$	$\frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\theta})Y(e^{j(\omega - \theta)})d\theta$	Modulation

## Common Fourier transform pairs

Sequence	Fourier transform
$\delta[n]$	1
$\delta[n - n_0]$	$e^{-j\omega n_0}$
1 ( $-\infty < n < \infty$ )	$\sum_{k=-\infty}^{\infty} 2\pi\delta(\omega + 2\pi k)$
$a^n u[n]$ ( $ a  < 1$ )	$\frac{1}{1 - ae^{-j\omega}}$
$u[n]$	$\frac{1}{1 - e^{-j\omega}} + \sum_{k=-\infty}^{\infty} \pi\delta(\omega + 2\pi k)$
$(n + 1)a^n u[n]$ ( $ a  < 1$ )	$\frac{1}{(1 - ae^{-j\omega})^2}$
$\frac{\sin(\omega_c n)}{\pi n}$	$X(e^{j\omega}) = \begin{cases} 1 &  \omega  < \omega_c \\ 0 & \omega_c <  \omega  \leq \pi \end{cases}$
$x[n] = \begin{cases} 1 & 0 \leq n \leq M \\ 0 & \text{otherwise} \end{cases}$	$\frac{\sin[\omega(M+1)/2]}{\sin(\omega/2)} e^{-j\omega M/2}$
$e^{j\omega_0 n}$	$\sum_{k=-\infty}^{\infty} 2\pi\delta(\omega - \omega_0 + 2\pi k)$

## Common z-transform pairs

Sequence	Transform	ROC
$\delta[n]$	1	All $z$
$u[n]$	$\frac{1}{1 - z^{-1}}$	$ z  > 1$
$-u[-n - 1]$	$\frac{1}{1 - z^{-1}}$	$ z  < 1$
$\delta[n - m]$	$z^{-m}$	All $z$ except 0 or $\infty$
$a^n u[n]$	$\frac{1}{1 - az^{-1}}$	$ z  >  a $
$-a^n u[-n - 1]$	$\frac{1}{1 - az^{-1}}$	$ z  <  a $
$na^n u[n]$	$\frac{az^{-1}}{(1 - az^{-1})^2}$	$ z  >  a $
$-na^n u[-n - 1]$	$\frac{az^{-1}}{(1 - az^{-1})^2}$	$ z  <  a $
$\begin{cases} a^n & 0 \leq n \leq N - 1, \\ 0 & \text{otherwise} \end{cases}$	$\frac{1 - a^N z^{-N}}{1 - az^{-1}}$	$ z  > 0$
$\cos(\omega_0 n) u[n]$	$\frac{1 - \cos(\omega_0)z^{-1}}{1 - 2\cos(\omega_0)z^{-1} + z^{-2}}$	$ z  > 1$
$r^n \cos(\omega_0 n) u[n]$	$\frac{1 - r \cos(\omega_0)z^{-1}}{1 - 2r \cos(\omega_0)z^{-1} + r^2 z^{-2}}$	$ z  > r$