

# EEE4001F: Digital Signal Processing

Class Test 1

16 March 2012

## SOLUTIONS

---

**Name:**

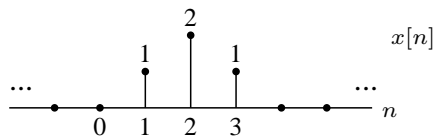
**Student number:**

---

### Information

- The test is closed-book.
  - This test has *four* questions, totalling 20 marks.
  - Answer *all* the questions.
  - You have 45 minutes.
-

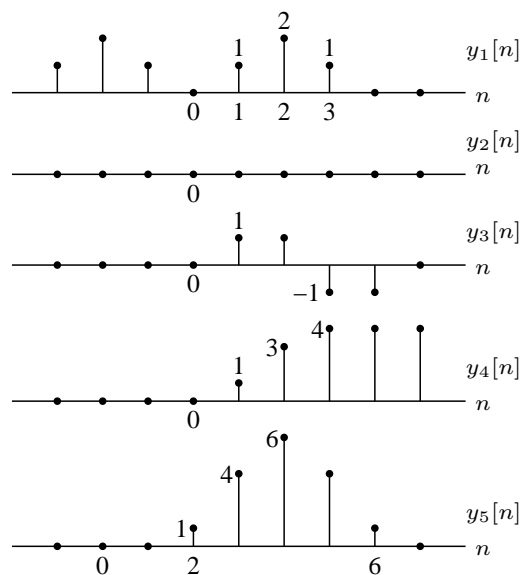
1. (5 marks) For the signal  $x[n]$  below



plot the following:

- (a)  $y_1[n] = x[n] + x[-n]$
- (b)  $y_2[n] = x[n]x[-n]$
- (c)  $y_3[n] = x[n] - x[n - 1]$
- (d)  $y_4[n] = \sum_{k=-\infty}^n x[k]$
- (e)  $y_5[n] = x[n] * x[n]$ .

Plots follow:



2. (5 marks) For what values of  $\omega$  is the signal  $x[n] = e^{j\omega n}$  periodic with a period of 8?

For  $x[n]$  to be periodic with period  $N$  we require that  $x[n] = x[n + N]$  for all  $n$ . For the given  $x[n]$  this requires that

$$e^{j\omega n} = e^{j\omega(n+N)} = e^{j\omega N} e^{j\omega n}$$

for all  $n$ , which means that  $e^{j\omega N} = 1 = e^{j2\pi k}$  for any integer  $k$ . Thus  $\omega$  must satisfy

$$\omega = \frac{2\pi k}{N} = \frac{2\pi}{8} k = \frac{\pi}{4} k$$

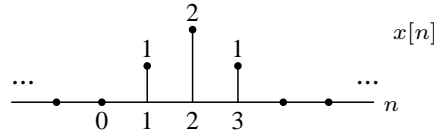
for any integer  $k$ . The set of frequencies  $\omega$  for which the signal is periodic with a period of 8 is therefore

$$\omega \in \left\{ \dots, -\frac{3\pi}{4}, -\frac{\pi}{2}, -\frac{\pi}{4}, 0, \frac{\pi}{4}, \frac{\pi}{2}, \frac{3\pi}{4}, \dots \right\}.$$

3. (5 marks) Suppose  $x[n]$  has the DTFT  $X(e^{j\omega})$ , and consider the signal  $y[n]$  defined as follows:

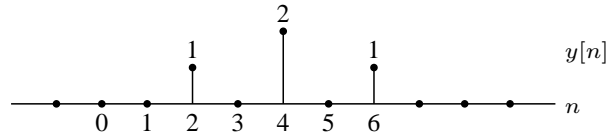
$$y[n] = \begin{cases} x[n/2] & n \text{ even} \\ 0 & n \text{ odd.} \end{cases}$$

- (a) Plot  $y[n]$  if  $x[n]$  is the following:



- (b) Find a general expression for the DTFT  $y(e^{j\omega})$  of  $y[n]$  in terms of  $X(e^{j\omega})$ .

- (a) The signal is expanded on the  $n$  axis as follows:



- (b) Since

$$\begin{aligned} Y(e^{j\omega}) &= \sum_{n=-\infty}^{\infty} y[n]e^{-j\omega n} \\ &= \dots + y[-3]e^{j\omega 3} + y[-2]e^{j\omega 2} + y[-1]e^{j\omega 1} + y[0]e^{j\omega 0} + y[1]e^{-j\omega 1} \\ &\quad + y[2]e^{-j\omega 2} + y[3]e^{-j\omega 3} + \dots \\ &= \dots + 0e^{j\omega 3} + x[-1]e^{j\omega 2} + 0e^{j\omega 1} + x[0]e^{j\omega 0} + 0e^{-j\omega 1} \\ &\quad + x[1]e^{-j\omega 2} + 0e^{-j\omega 3} + \dots \\ &= \dots + x[-1]e^{j\omega 2} + x[0]e^{j\omega 0} + x[1]e^{-j\omega 2} + \dots = \sum_{n=-\infty}^{\infty} x[n]e^{-j(2\omega)n} \end{aligned}$$

we evidently have  $Y(e^{j\omega}) = X(e^{j(2\omega)})$ , which is a compression of the signal on the  $\omega$  axis.

4. (5 marks) A linear time-invariant system has system function

$$H(z) = \frac{6z - 2}{6z - 3}, \quad |z| > 1/2.$$

Find the impulse response of a stable inverse system. Is this inverse system causal?

The system function can be written as

$$H(z) = \frac{z - 1/3}{z - 1/2} = \frac{1 - 1/3z^{-1}}{1 - 1/2z^{-1}}$$

so the inverse is

$$G(z) = \frac{z - 1/2}{z - 1/3} = \frac{1 - 1/2z^{-1}}{1 - 1/3z^{-1}}.$$

This has a pole at  $z = 1/3$ , so the two possible ROCs are  $|z| < 1/3$  and  $|z| > 1/3$ . If  $|z| > 1/3$  then the inverse system is stable (ROC contains the unit circle) and it will also be causal (ROC extends out to infinity).

The inverse in this case is the right-sided sequence

$$h[n] = (1/3)^n u[n] - \frac{1}{2}(1/3)^{n-1} u[n-1].$$

## Fourier transform properties

Sequences $x[n], y[n]$	Transforms $X(e^{j\omega}), Y(e^{j\omega})$	Property
$ax[n] + by[n]$	$aX(e^{j\omega}) + bY(e^{j\omega})$	Linearity
$x[n - n_d]$	$e^{-j\omega n_d} X(e^{j\omega})$	Time shift
$e^{j\omega_0 n} x[n]$	$X(e^{j(\omega - \omega_0)})$	Frequency shift
$x[-n]$	$X(e^{-j\omega})$	Time reversal
$nx[n]$	$j \frac{dX(e^{j\omega})}{d\omega}$	Frequency diff.
$x[n] * y[n]$	$X(e^{-j\omega})Y(e^{-j\omega})$	Convolution
$x[n]y[n]$	$\frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\theta})Y(e^{j(\omega - \theta)})d\theta$	Modulation

## Common Fourier transform pairs

Sequence	Fourier transform
$\delta[n]$	1
$\delta[n - n_0]$	$e^{-j\omega n_0}$
1 $(-\infty < n < \infty)$	$\sum_{k=-\infty}^{\infty} 2\pi\delta(\omega + 2\pi k)$
$a^n u[n]$ $( a  < 1)$	$\frac{1}{1 - ae^{-j\omega}}$
$u[n]$	$\frac{1}{1 - e^{-j\omega}} + \sum_{k=-\infty}^{\infty} \pi\delta(\omega + 2\pi k)$
$(n + 1)a^n u[n]$ $( a  < 1)$	$\frac{1}{(1 - ae^{-j\omega})^2}$
$\frac{\sin(\omega_c n)}{\pi n}$	$X(e^{j\omega}) = \begin{cases} 1 &  \omega  < \omega_c \\ 0 & \omega_c <  \omega  \leq \pi \end{cases}$
$x[n] = \begin{cases} 1 & 0 \leq n \leq M \\ 0 & \text{otherwise} \end{cases}$	$\frac{\sin[\omega(M+1)/2]}{\sin(\omega/2)} e^{-j\omega M/2}$
$e^{j\omega_0 n}$	$\sum_{k=-\infty}^{\infty} 2\pi\delta(\omega - \omega_0 + 2\pi k)$

## Common z-transform pairs

Sequence	Transform	ROC
$\delta[n]$	1	All $z$
$u[n]$	$\frac{1}{1 - z^{-1}}$	$ z  > 1$
$-u[-n - 1]$	$\frac{1}{1 - z^{-1}}$	$ z  < 1$
$\delta[n - m]$	$z^{-m}$	All $z$ except 0 or $\infty$
$a^n u[n]$	$\frac{1}{1 - az^{-1}}$	$ z  >  a $
$-a^n u[-n - 1]$	$\frac{1}{1 - az^{-1}}$	$ z  <  a $
$na^n u[n]$	$\frac{az^{-1}}{(1 - az^{-1})^2}$	$ z  >  a $
$-na^n u[-n - 1]$	$\frac{az^{-1}}{(1 - az^{-1})^2}$	$ z  <  a $
$\begin{cases} a^n & 0 \leq n \leq N - 1, \\ 0 & \text{otherwise} \end{cases}$	$\frac{1 - a^N z^{-N}}{1 - az^{-1}}$	$ z  > 0$
$\cos(\omega_0 n)u[n]$	$\frac{1 - \cos(\omega_0)z^{-1}}{1 - 2\cos(\omega_0)z^{-1} + z^{-2}}$	$ z  > 1$
$r^n \cos(\omega_0 n)u[n]$	$\frac{1 - r\cos(\omega_0)z^{-1}}{1 - 2r\cos(\omega_0)z^{-1} + r^2 z^{-2}}$	$ z  > r$