

EEE401F Class Test

30 April 2003

Name:

Student number:

Information

- The test is closed-book.
 - This test has *six* questions, totalling 30 marks.
 - Answer *all* the questions.
 - You have 45 minutes.
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1. (5 marks) Consider the system function

$$H(z) = \frac{z - 1}{(z - 2)^2}$$

for a *causal* system. Draw a pole-zero diagram for the system, and indicate the region of convergence. Is the system stable? Why?

- Single zero at $z = 1$ since $H(z)|_{z=1} = 0$
- Double pole at $z = 2$ since $H(z)|_{z=2} = \infty$
- The system is causal, so the ROC is outside of the outermost pole: $|z| > 2$
- The system is stable if the ROC includes the unit circle, which it does not. Hence the system is unstable.

2. (5 marks) Consider the system function

$$H(z) = \frac{z - 1}{(z - 2)^2}$$

for a *causal* system. Using the power series method, find the first five values of the causal discrete-time impulse response of the system.

Start with long division of $H(z)$ to get the required power series expansion:

$$\begin{array}{r} z^{-1} + 3z^{-2} + 8z^{-3} + 20z^{-4} + \dots \\ z^2 - 4z + 4 \overline{) \quad z \quad -1} \\ \underline{z \quad -4 + 4z^{-1}} \\ 3 \quad -4z^{-1} \\ \underline{3 - 12z^{-1} + 12z^{-2}} \\ 8z^{-1} - 12z^{-2} \\ \underline{8z^{-1} - 32z^{-2} + 32z^{-3}} \\ 20z^{-2} - 32z^{-3} \end{array}$$

Now, since

$$\begin{aligned} H(z) &= \sum_{n=-\infty}^{\infty} h[n]z^{-n} \\ &= \dots + h[-2]z^2 + h[-1]z^1 + h[0] + h[1]z^{-1} + h[2]z^{-2} + \dots, \end{aligned}$$

we can simply equate coefficients to get the impulse response values

$$h[n] = 0 \quad \text{for } n < 0$$

$$h[0] = 0$$

$$h[1] = 1$$

$$h[2] = 3$$

$$h[3] = 8$$

$$h[4] = 20$$

$$h[5] = \dots$$

3. (6 marks) Consider the system function

$$H(z) = \frac{z-1}{(z-2)^2}$$

for a *causal* system. Using the z-transform pair

$$na^n u[n] \xleftrightarrow{z} \frac{az^{-1}}{(1-az^{-1})^2}, \quad |z| > |a|$$

and the property

$$x[n-n_0] \xleftrightarrow{z} z^{-n_0} X(z),$$

find a closed-form expression for the impulse response $h[n]$.

Write the system function in terms of quantities that we know the inverse transform of:

$$\begin{aligned} H(z) &= \frac{z^{-2}(z-1)}{[z^{-1}(z-2)]^2} = \frac{z^{-1}-z^{-2}}{(1-2z^{-1})^2} \\ &= \frac{1}{2} \frac{2z^{-1}}{(1-2z^{-1})^2} - \frac{z^{-1}}{2} \frac{2z^{-1}}{(1-2z^{-1})^2}, \end{aligned}$$

so

$$h[n] = \frac{1}{2}n2^n u[n] - \frac{1}{2}(n-1)2^{n-1}u[n-1].$$

Check by evaluating the first 5 points:

$$h[0] = \frac{1}{2}(0) = 0$$

$$h[1] = \frac{1}{2}(2) - 0 = 1$$

$$h[2] = \frac{1}{2}(2)(4) - \frac{1}{2}(2) = 4 - 1 = 3$$

$$h[3] = \frac{1}{2}(3)(2^3) - \frac{1}{2}(2)(2^2) = 12 - 4 = 8$$

$$h[4] = \dots$$

4. (4 marks) Consider the system function

$$H(z) = \frac{z - 1}{(z - 2)^2}$$

for a *causal* system. Find a difference equation (relating the input $x[n]$ to the output $y[n]$) for the system.

$$H(z) = \frac{Y(z)}{X(z)} = \frac{z - 1}{z^2 - 4z + 4},$$

so

$$z^2 Y(z) - 4z Y(z) + 4Y(z) = z X(z) - X(z).$$

Inverting gives

$$y[n + 2] - 4y[n + 1] + 4y[n] = x[n + 1] - x[n].$$

Alternatively,

$$H(z) = \frac{Y(z)}{X(z)} = \frac{z^{-1} - z^{-2}}{1 - 4z^{-1} + 4z^{-2}},$$

so

$$Y(z) - 4z^{-1} Y(z) + 4z^{-2} Y(z) = z^{-1} X(z) - z^{-2} X(z).$$

Inverting gives

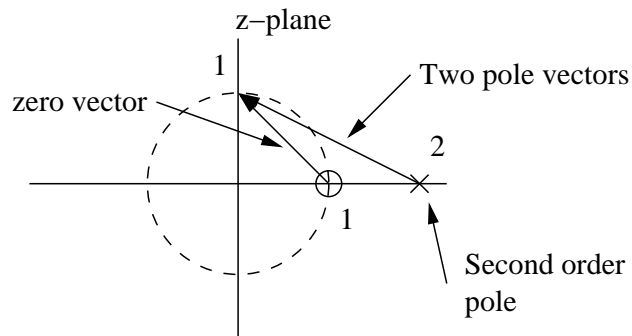
$$y[n] - 4y[n - 1] + 4y[n - 2] = x[n - 1] - x[n - 2].$$

5. (6 marks) Consider the system function

$$H(z) = \frac{z - 1}{(z - 2)^2}$$

for a *causal* system. Use graphical arguments to find the magnitude and phase of the frequency response of the system for $\omega = \pi/2$ and $\omega = \pi$.

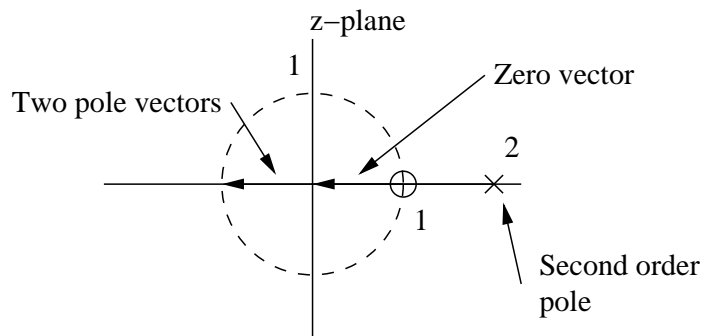
For $\omega = \pi/2$:



$$\text{Gain} = \frac{\prod \text{length of zero vectors}}{\prod \text{length of pole vectors}} = \frac{\sqrt{1^2 + 1^2}}{\sqrt{1^2 + 2^2} \sqrt{1^2 + 2^2}} = \frac{\sqrt{2}}{5}.$$

$$\begin{aligned} \text{Phase} &= \sum \text{angle of zero vectors} - \sum \text{angle of pole vectors} \\ &= \frac{3\pi}{4} - 2(\pi - \tan^{-1}(1/2)) = \frac{3\pi}{4} - 2(2.6779) = -2.9997 \text{ rad} = -171.87^\circ = 188.13^\circ. \end{aligned}$$

For $\omega = \pi$:



$$\text{Gain} = \frac{\prod \text{length of zero vectors}}{\prod \text{length of pole vectors}} = \frac{2}{(3)(3)} = \frac{2}{9}.$$

$$\begin{aligned} \text{Phase} &= \sum \text{angle of zero vectors} - \sum \text{angle of pole vectors} \\ &= \pi - 2(\pi) = -\pi \text{ rad} = \pi \text{ rad} = 180^\circ. \end{aligned}$$

6. (4 marks) Consider the system function

$$H(z) = \frac{z - 1}{(z - 2)^2}$$

for a *causal* system. Use $H(z)$ directly (algebraically) to find the magnitude and phase of the frequency response of the system for $\omega = \pi/2$.

We want to know the magnitude and phase of

$$H(e^{j\pi/2}) = \left. \frac{z - 1}{(z - 2)^2} \right|_{z=e^{j\pi/2}} = \frac{e^{j\pi/2} - 1}{(e^{j\pi/2} - 2)^2} = \frac{j - 1}{(j - 2)^2}$$

(since $e^{j\pi/2} = j$). Simplifying gives

$$\begin{aligned} H(e^{j\pi/2}) &= \frac{i - 1}{i^2 - 4i + 4} = \frac{i - 1}{3 - 4i} = \frac{i - 1}{3 - 4i} \frac{3 + 4i}{3 + 4i} \\ &= \frac{3i - 4 - 3 - 4i}{9 + 16} = \frac{-7 - j}{25}. \end{aligned}$$

The magnitude of the response is therefore

$$|H(e^{j\pi/2})| = \left| \frac{-7 - j}{25} \right| = \frac{\sqrt{49 + 1}}{25} = \frac{\sqrt{25(2)}}{5(5)} = \frac{\sqrt{2}}{5}.$$

The phase is the angle of this vector, which lies in the third quadrant of the plane. The principal angle is therefore

$$\theta_p = \tan^{-1} \left(\frac{1}{7} \right) = 8.13^\circ,$$

and the overall phase is $180^\circ + \theta_p = 188.13^\circ$.