EEE2035F EXAM SIGNALS AND SYSTEMS I HINTS: June 2011

1. Most of these are routine. For part (c), note that the sifting property implies that

$$y_3(t) = \int_{-\infty}^{\infty} x(t+1)\delta(t-\lambda)d\lambda = x(t+1).$$

For part (d), because the signal is zero for negative time, the signal $y_4(t)$ is just the indefinite integral of x(t) (i.e. the lower limit of zero can be changed to $-\infty$ without modifying the result. For part (e) you can plot the signal x(t-1) and then take the generalised derivative (including the impulse in the result), or alternatively just take the derivative of x(t) and shift — the order of operations doesn't matter because shifting and differentiating are both linear operations.

- 2. For the first two parts substitute and use the sifting property to find (a) y(t) = g(t-2)and (b) y(t) = g(t-4). The two inputs differ by a shift of one unit, while the outputs differ by a shift of two, so the system is not time invariant (and therefore not LTI).
- 3. The output is y(t) = h(t) * x(t). Using the derivative property y'(t) = h'(t) * x(t) (where prime denotes derivative). The signal h'(t) is just two delta functions, so the convolution is simple and one can easily find y'(t). Integrate (indefinite) to get the required result.
- 4. You can use first principles to calculate the Fourier series coefficients, although it is tricky to get the result into a symmetric form where the magnitude and phase can be easily obtained. More simply, using the given series representation along with the observed relationship y(t) = 2x(t 1/2) a useful form for the coefficients can be obtained directly.

- 5. The frequency response is the Fourier transform of the impulse response, and can be seen to be $H(\omega) = 2\pi p_3(\omega)$. This is an ideal lowpass filter that passes frequencies $|\omega| \leq 3/2$ (with a gain of 2π) and eliminates the rest. Thus the response to $x_1(t)$ will be $2\pi \cos(t)$, and the response to $x_2(t)$ will be zero. These results can be formally shown by noting that a complex exponential $e^{j\omega_0 t}$ into a linear system produces the output $H(\omega_0)e^{j\omega_0 t}$. Else consider plots of the signals in the frequency domain use multiplication to get the required results.
- 6. The input-output relationship for a linear system in frequency coordinates is $Y(\omega) = H(\omega)X(\omega)$, so the frequency response is $H(\omega) = Y(\omega)/X(\omega)$ after finding $Y(\omega)$ and $X(\omega)$. The impulse response is the inverse Fourier transform of $H(\omega)$, which can be found using a partial fraction expansion.
- 7. The result for (a) can be found by applying the time shift property to the signal $e^{-3t}u(t)$ and manipulating accordingly. For part (b), ignore the differentiation operation initially, and use the Euler expansion for sin to find the transform by applying the frequency translation property to the two parts of the signal. Differentiation in time is then just multiplication by $j\omega$.
- Multiplication by a cosine corresponds to modulation or frequency domain shifting. The signal V(ω) contains two replicas of X(ω), one at ω = 4 and one at ω = -4, both with height π. The signal z(t) is the same operation but instead applied to V(ω), and contains two replicas of X(ω) (each of height π²) at ω = ±8, and one of height 2π² at DC (ω = 0). To reconstruct x(t) at the output the filter should pass this central replica and scale it to have unit height. The cutoff is variable but must be in the range 2 ≤ ω ≤ 6.