

# EEE2035F: Signals and Systems I

## Class Test 2

19 May 2006

Name:

Student number:

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### Information

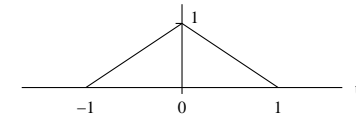
- The test is closed-book.
  - This test has *four* questions, totalling 20 marks.
  - There is an information sheet attached at the end of this paper.
  - Answer *all* the questions.
  - You have 45 minutes.
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1. (5 marks) Sketch the signal

$$x(t) = \begin{cases} t + 1 & -1 \leq t \leq 0 \\ -t + 1 & 0 \leq t \leq 1 \end{cases}$$

and find its Fourier transform.

Sketch follows:



Letting  $y(t) = dx(t)/dt$  we have  $y(t) = p_1(t + 1/2) - p_1(t - 1/2)$ , which has Fourier transform

$$Y(\omega) = \text{sinc}(\omega/2\pi)e^{j\omega/2} - \text{sinc}(\omega/2\pi)e^{-j\omega/2}.$$

From fundamental calculus the inverse of the derivative  $y(t) = dx(t)/dt$  is

$$x(t) = \int_{-\infty}^t y(\lambda)d\lambda + c.$$

Since  $\lim_{t \rightarrow -\infty} x(t) = 0$  for the given signal we must have

$$\lim_{t \rightarrow -\infty} \int_{-\infty}^t y(\lambda)d\lambda + c = 0,$$

so  $c = 0$  (which is true for most signals we care about). Thus the left hand side of the expression for the integration property

$$\int_{-\infty}^t y(\lambda)d\lambda \leftrightarrow \frac{1}{j\omega}Y(\omega) + \pi Y(0)\delta(\omega)$$

is the signal  $x(t)$ , and the right hand side is the required Fourier transform  $X(\omega)$ . Since  $Y(0) = \lim_{\omega \rightarrow 0} = 1 - 1 = 0$ , the solution is

$$\begin{aligned} X(\omega) &= \frac{1}{j\omega}Y(\omega) = \frac{1}{j\omega} \text{sinc}\left(\frac{\omega}{2\pi}\right) \frac{2j}{2j} (e^{j\omega/2} - e^{-j\omega/2}) \\ &= \frac{2j}{j\omega} \text{sinc}\left(\frac{\omega}{2\pi}\right) \sin(\omega/2) = \text{sinc}\left(\frac{\omega}{2\pi}\right) \text{sinc}\left(\frac{\omega}{2\pi}\right) = \text{sinc}^2\left(\frac{\omega}{2\pi}\right). \end{aligned}$$

Note also that we can write  $x(t)$  as  $(1 - |t|)p_2(t)$ , which is in the table of transform pairs. Alternatively, observe that  $x(t) = p_1(t) * p_1(t)$  and the required transform follows immediately.

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2. (5 marks) Find  $y(t) = x(t) * x(t)$  with

$$x(t) = 2\text{sinc}\left(\frac{t}{4\pi}\right).$$

(Note that reasoning in the frequency domain makes this question much simpler.)

Since convolution in time domain corresponds to multiplication in the frequency domain, the relationship between the signals is  $Y(\omega) = X(\omega)X(\omega)$ . From the Fourier transform tables

$$\frac{1}{2}\text{sinc}\left(\frac{t}{4\pi}\right) \iff 2\pi p_{1/2}(\omega)$$

so

$$X(\omega) = 8\pi p_{1/2}(\omega)$$

and

$$Y(\omega) = X(\omega)X(\omega) = (8\pi)^2 p_{1/2}(\omega)p_{1/2}(\omega) = 32\pi[2\pi p_{1/2}(\omega)].$$

The inverse Fourier transform is obtained from the tables as

$$y(t) = 16\pi\text{sinc}\left(\frac{t}{4\pi}\right).$$

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3. (5 marks) Find the inverse Fourier transform of the signal

$$X(\omega) = e^{j5\omega} \frac{1}{4 + j\omega}.$$

Starting with the FT pair

$$x_1(t) = e^{-4t}u(t) \iff X_1(\omega) = \frac{1}{4 + j\omega}$$

and applying the time shift property gives

$$x_1(t + 5) = e^{-4(t+5)}u(t + 5) \iff X_1(\omega)e^{j\omega 5} = \frac{1}{4 + j\omega}e^{j\omega 5}.$$

Thus the required inverse transform is

$$x(t) = e^{-4(t+5)}u(t + 5).$$

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4. (5 marks) Sketch each of the following transfer functions and classify them as having a lowpass, highpass, bandpass, or bandstop frequency response.

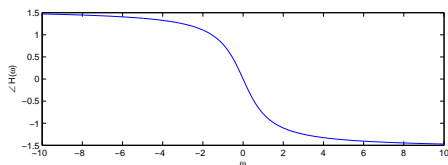
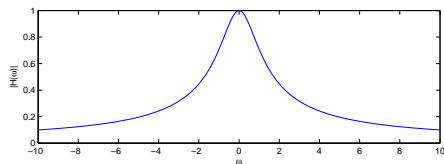
(a)  $H(\omega) = \frac{1}{1+j\omega}$

(b)  $H(\omega) = \frac{j\omega}{1+j\omega}$ .

(a) The magnitude response for  $H(\omega) = \frac{1}{1+j\omega}$  satisfies

$$|H(\omega)|^2 = H(\omega)H^*(\omega) = \frac{1}{1+j\omega} \frac{1}{1-j\omega} = \frac{1}{1+\omega^2}.$$

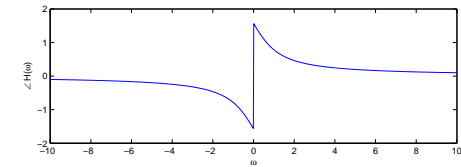
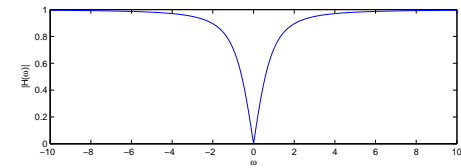
Thus  $|H(\omega)|^2 = 1$  and  $|H(\omega)|^2 \rightarrow 0$  as  $\omega \rightarrow \infty$ , so the filter has a lowpass characteristic.



(b) The magnitude response for  $H(\omega) = \frac{j\omega}{1+j\omega}$  is

$$|H(\omega)|^2 = H(\omega)H^*(\omega) = \frac{j\omega}{1+j\omega} \frac{-j\omega}{1-j\omega} = \frac{\omega^2}{1+\omega^2}.$$

Now  $|H(\omega)|^2 = 0$  and  $|H(\omega)|^2 \rightarrow 1$  as  $\omega \rightarrow \infty$ , so the filter has a highpass characteristic.



## INFORMATION SHEET

### Fourier transform properties

Property	Transform Pair/Property
Linearity	$ax(t) + bv(t) \leftrightarrow aX(\omega) + bV(\omega)$
Time shift	$x(t - c) \leftrightarrow X(\omega)e^{-j\omega c}$
Time scaling	$x(at) \leftrightarrow \frac{1}{ a }X(\frac{\omega}{a}) \quad a > 0$
Time reversal	$x(-t) \leftrightarrow X(-\omega) = \overline{X(\omega)}$
Multiplication by power of $t$	$t^n x(t) \leftrightarrow j^n \frac{d^n}{d\omega^n} X(\omega) \quad n = 1, 2, \dots$
Multiplication by complex exponential	$x(t)e^{j\omega_0 t} \leftrightarrow X(\omega - \omega_0) \quad \omega_0 \text{ real}$
Multiplication by $\cos(\omega_0 t)$	$x(t)\cos(\omega_0 t) \leftrightarrow \frac{1}{2}[X(\omega + \omega_0) + X(\omega - \omega_0)]$
Differentiation in time domain	$\frac{d^n}{dt^n} x(t) \leftrightarrow (j\omega)^n X(\omega) \quad n = 1, 2, \dots$
Integration	$\int_{-\infty}^t x(\lambda)d\lambda \leftrightarrow \frac{1}{j\omega} X(\omega) + \pi X(0)\delta(\omega)$
Convolution in time domain	$x(t) * v(t) \leftrightarrow X(\omega)V(\omega)$
Multiplication in time domain	$x(t)v(t) \leftrightarrow \frac{1}{2\pi} X(\omega) * V(\omega)$
Parseval's theorem	$\int_{-\infty}^{\infty} x(t)v(t)dt = \frac{1}{2\pi} \int_{-\infty}^{\infty} \overline{X(\omega)}V(\omega)d\omega$
Parseval's theorem (special case)	$\int_{-\infty}^{\infty} x^2(t)dt = \frac{1}{2\pi} \int_{-\infty}^{\infty}  X(\omega) ^2 d\omega$
Duality	$X(t) \leftrightarrow 2\pi x(-\omega)$

### Common Fourier Transform Pairs

$x(t)$	$X(\omega)$
1 $(-\infty < t < \infty)$	$2\pi\delta(\omega)$
$-0.5 + u(t)$	$\frac{1}{j\omega}$
$u(t)$	$\pi\delta(\omega) + \frac{1}{j\omega}$
$\delta(t)$	1
$\delta(t - c)$	$e^{-j\omega c} \quad (c \text{ any real number})$
$e^{-bt}u(t)$	$\frac{1}{j\omega + b} \quad (b > 0)$
$e^{j\omega_0 t}$	$2\pi\delta(\omega - \omega_0) \quad (\omega_0 \text{ any real number})$
$p_\tau(t)$	$\tau \text{sinc} \frac{\omega\tau}{2\pi}$
$\tau \text{sinc} \frac{\omega\tau}{2\pi}$	$2\pi p_\tau(\omega)$
$(1 - \frac{2 t }{\tau})p_\tau(t)$	$\frac{\tau}{2} \text{sinc}^2(\frac{\omega\tau}{4\pi})$
$\frac{\tau}{2} \text{sinc}^2 \frac{\omega\tau}{4\pi}$	$2\pi(1 - \frac{2 \omega\tau }{\tau})p_\tau(\omega)$
$\cos(\omega_0 t + \theta)$	$\pi[e^{-j\theta}\delta(\omega + \omega_0) + e^{j\theta}\delta(\omega - \omega_0)]$
$\sin(\omega_0 t + \theta)$	$j\pi[e^{-j\theta}\delta(\omega + \omega_0) - e^{j\theta}\delta(\omega - \omega_0)]$

### Trigonometric identities

$$\sin(-\theta) = -\sin(\theta) \quad \cos(-\theta) = \cos(\theta) \quad \tan(-\theta) = -\tan(\theta)$$

$$\sin^2(\theta) + \cos^2(\theta) = 1 \quad \sin(2\theta) = 2\sin(\theta)\cos(\theta)$$

$$\cos(2\theta) = \cos^2(\theta) - \sin^2(\theta) = 2\cos^2(\theta) - 1 = 1 - 2\sin^2(\theta)$$

$$\sin(\theta_1 + \theta_2) = \sin(\theta_1)\cos(\theta_2) + \cos(\theta_1)\sin(\theta_2) \quad \cos(\theta_1 + \theta_2) = \cos(\theta_1)\cos(\theta_2) - \sin(\theta_1)\sin(\theta_2)$$

$$e^{j\theta} = \cos(\theta) + j\sin(\theta)$$